

# 1.6 元素訊號( Elementary signals )

- ❖ 指數訊號(exponential signals)
- ❖ 弦波訊號(sinusoidal signals)
- ❖ 弦波訊號與複數指數訊號
- ❖ 指數衰退之弦波訊號
- ❖ 步階(Step)函數(訊號)
- ❖ 脈衝(impulse)函數(訊號)
  - 微分
- ❖ RAMP function

# 指數訊號(exponential signals)

- ❖ (a) 自然界中常見之衰退或成長函數, 如:  
人口自然成長函數, 放射物質之衰退
- ❖ (b) 基本函數型態

$$x(t) = Be^{at},$$

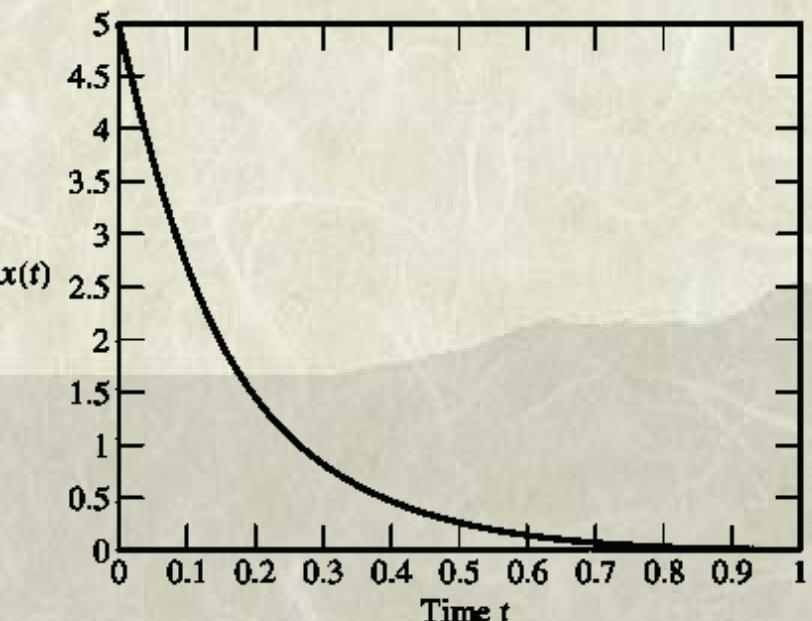
$B$ :  $t = 0$ 之初值

$a$ : 指數蛻變常數

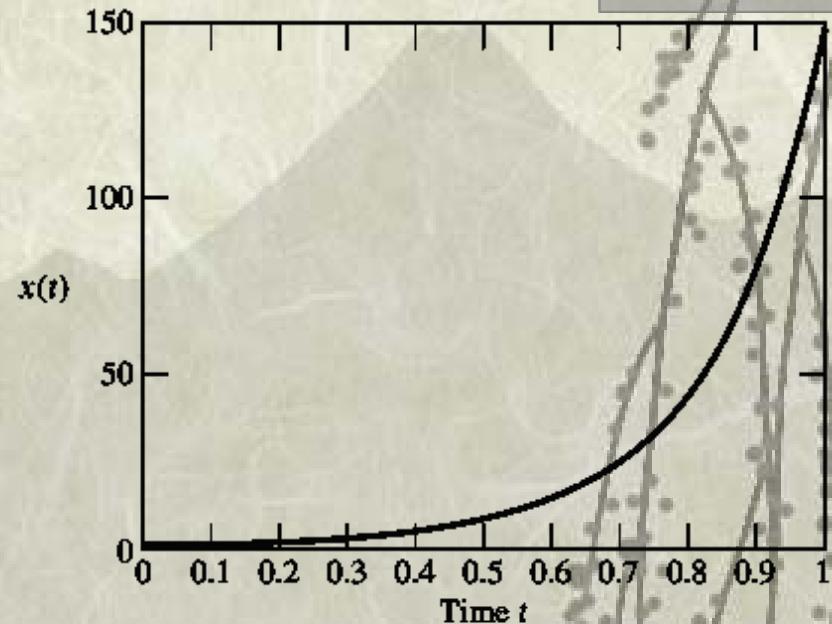
$a > 0$ , 成長

$a < 0$ , 衰退

# 圖例: 指數訊號(p34, F1.28)

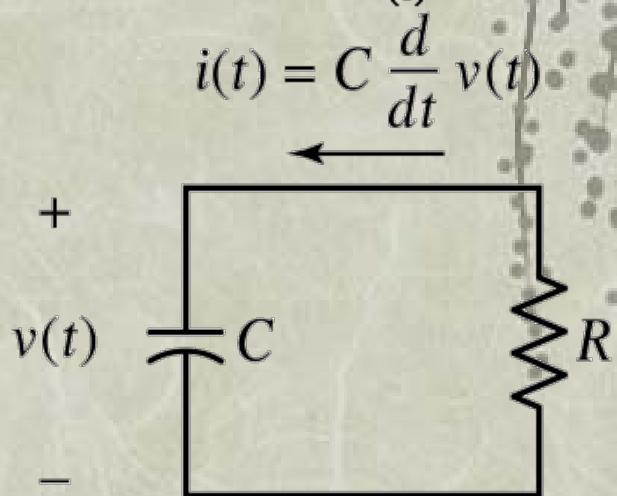


(a)



(b)

- ❖ (a)  $x(t) = Be^{-t/\tau}, \tau > 0$
- ❖ (b)  $x(t) = Be^{t/\tau}, \tau > 0$



# 計算: 一階 $RC$ 電路

$$Ri(t) + v(t) = 0$$

$$RC \frac{d}{dt} v(t) + v(t) = 0$$

(一階常微分方程式)

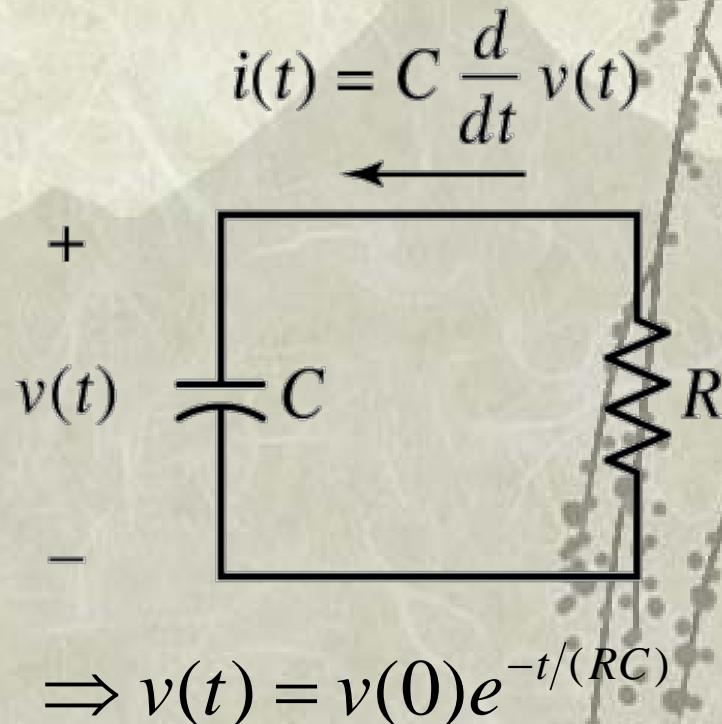
$$\text{假設 } v(t) = Be^{at}$$

$$RC \frac{d}{dt} v(t) + v(t) = 0$$

$$RCaBe^{at} + Be^{at} = 0$$

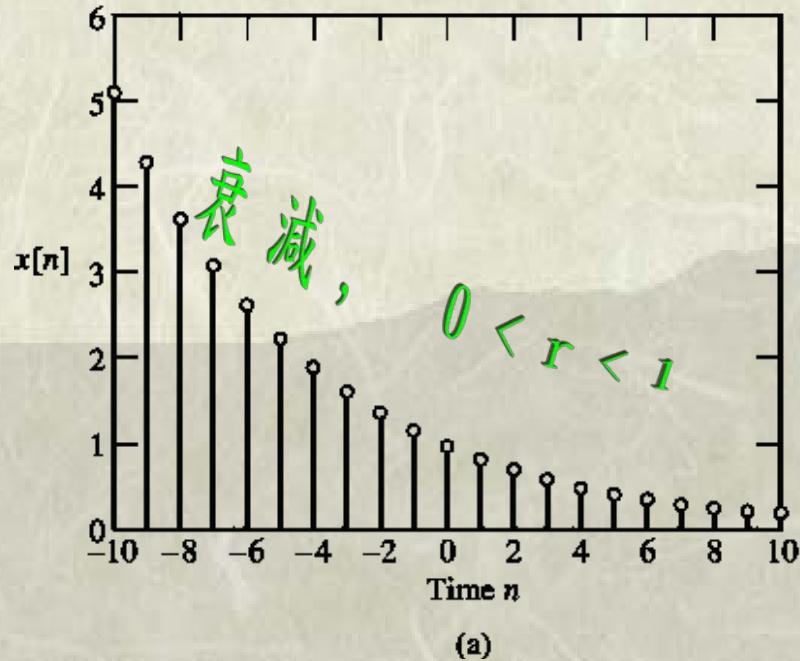
$$\rightarrow RCa + 1 = 0, \quad \therefore a = -1/(RC)$$

$$\rightarrow v(0) = Be^{a_0}, \quad \therefore B = v(0)$$

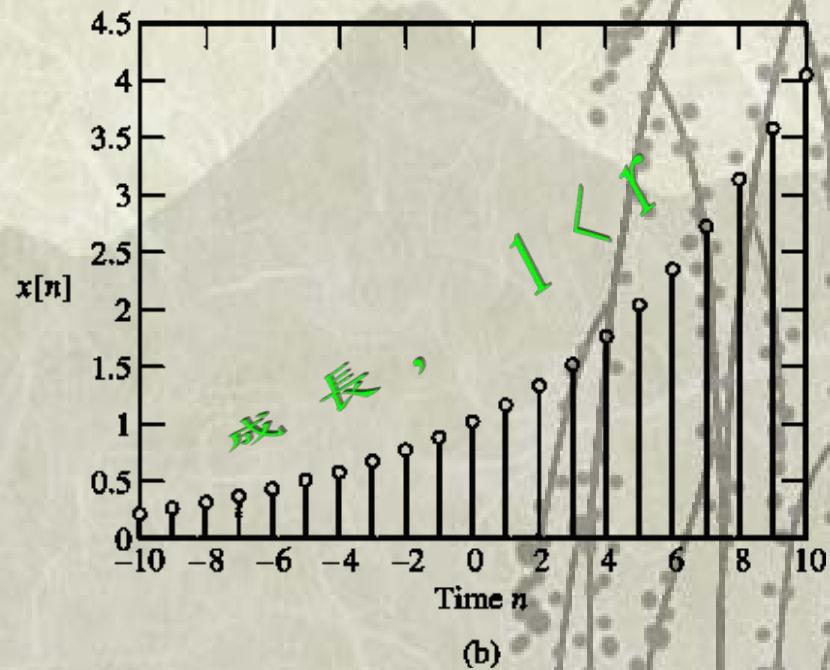


# 離散指數訊號(exponential signals)

ee: P35



(a)



(b)

- ❖ (a)  $x[n] = Br^n, 0 < r < 1$
- ❖ (b)  $x[n] = Br^n, r > 1$
- ❖ (c) 類比於連續指數  $r = e^\alpha$

# 複數指數訊號

$$x(t) = Be^{at}$$

$$x[n] = Br^n$$

- ❖ 若  $B, r, a$  為複數(complex) 則可定義出複數指數訊號
- ❖ 常見之複數指數訊號

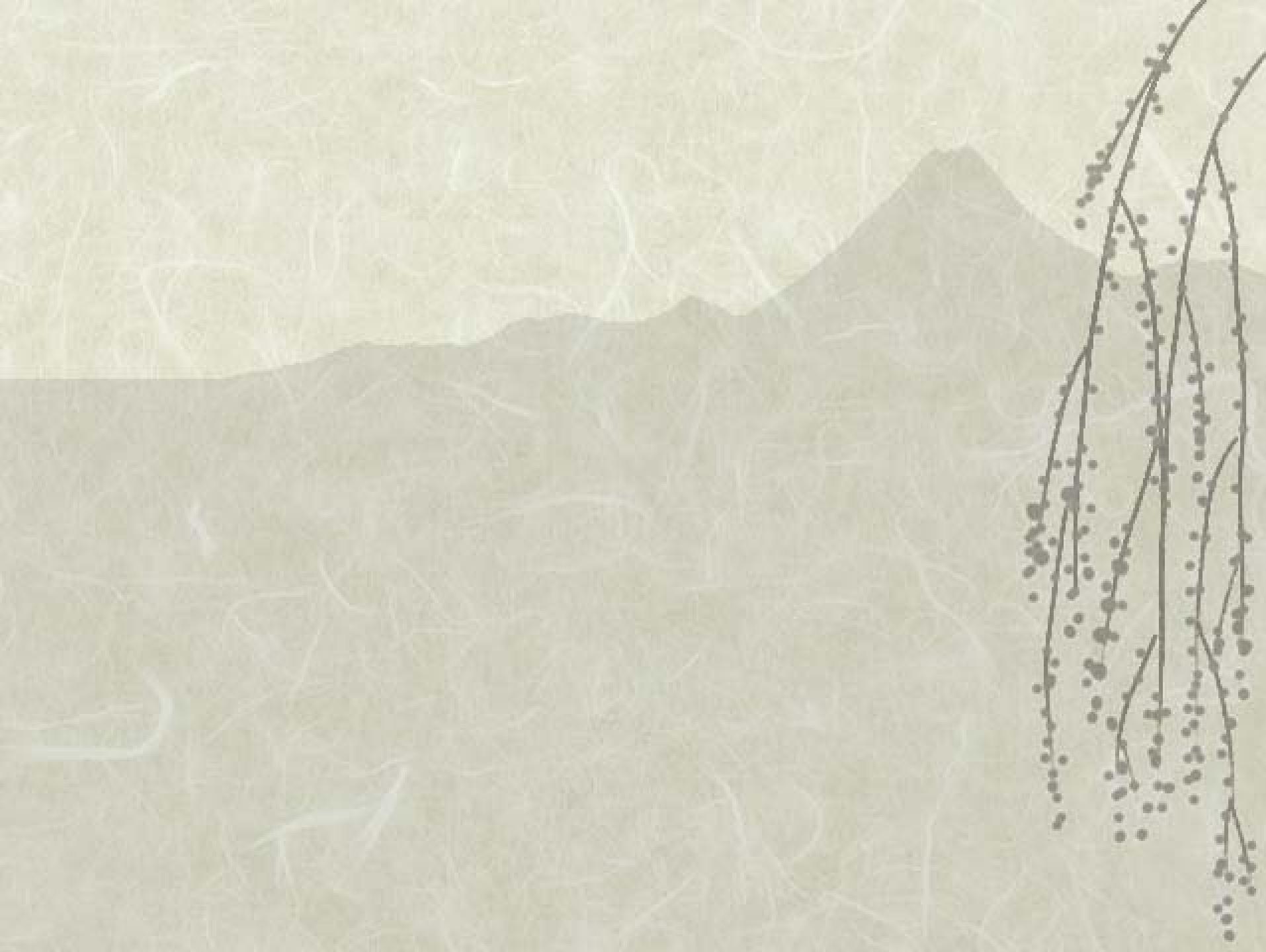
$$x(t) = Be^{j\omega t}$$

$$x[n] = Be^{j\Omega n} = B(e^{j\Omega})^n$$

Fourier transform 之  
基本基底訊號

←連續

←離散



# 弦波訊號(sinusoidal signals)

$$x(t) = A \cos(\varpi t + \phi) \quad (1.35)$$

重要參數

$A$ : 振幅

$\varpi$ : 角頻率

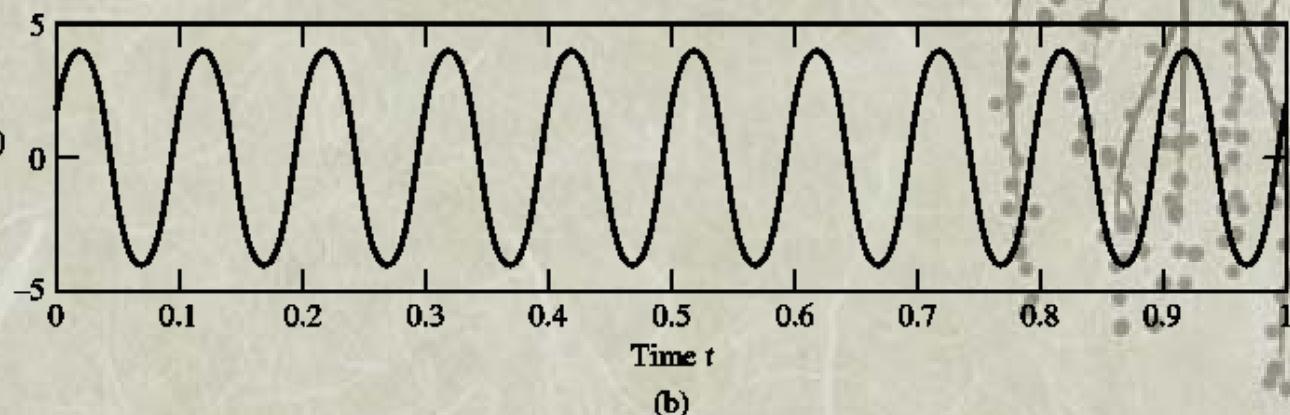
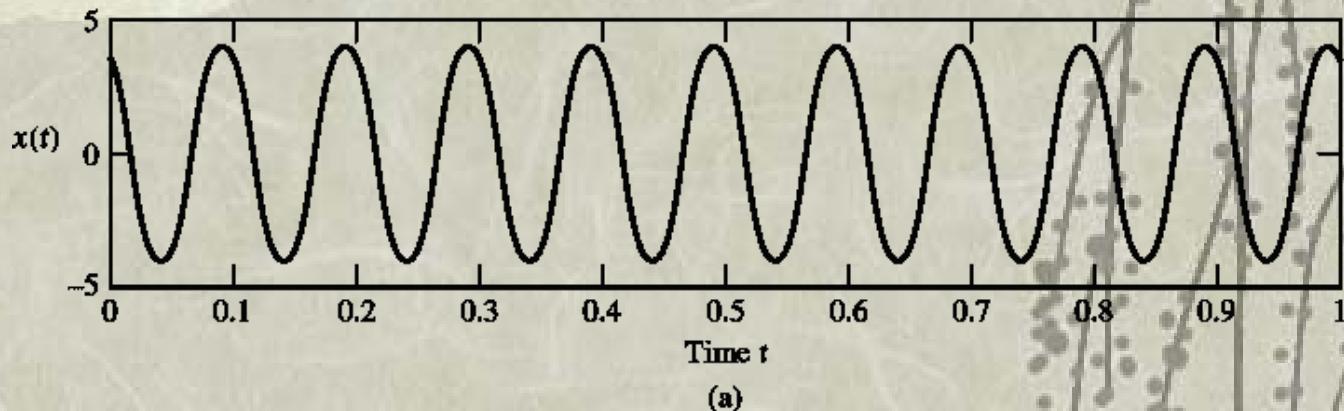
$$\varpi = 2\pi f$$

$\phi$ : 啟始相位

$$x(t+T) = x(t)$$

$$\varpi T = n2\pi, n \in \mathbf{Z}$$

$$\rightarrow T = \frac{2\pi}{\varpi}$$



計算: 二階LC電路

$$v(t) + \frac{1}{C} \int i(t) dt = 0$$

$$L \frac{d}{dt} i(t) + \frac{1}{C} \int i(t) dt = 0$$

$$LC \frac{d^2}{dt^2} v(t) + v(t) = 0$$

二階常微分方程式

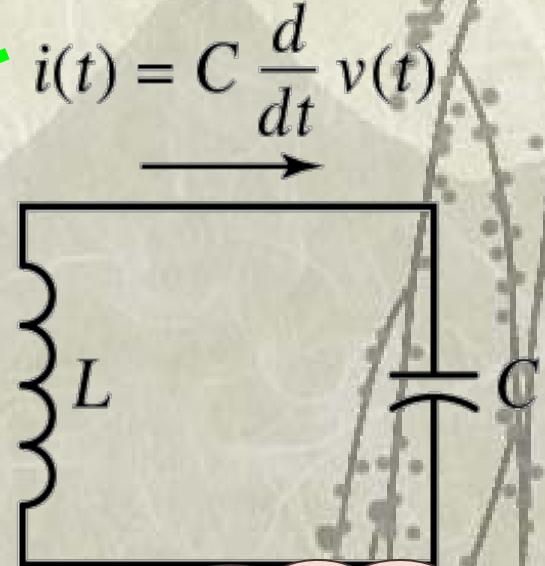
假設  $v(t) = Be^{at}$

$$LCa^2 Be^{at} + Be^{at} = 0$$

$$\rightarrow RCa^2 + 1 = 0, \quad \therefore a = \sqrt{-1/(RL)} = j\varpi_0.$$

$$\rightarrow \varpi_0 = \frac{1}{\sqrt{LC}}, \quad a \text{為複數, 解型態 } A_r \cos(\varpi_0 t) + jA_i \sin(\varpi_0 t)$$

$$\text{代入初值 } v(0), i(0) = 0 \Rightarrow v(t) = v(0) \cos(\varpi_0 t)$$

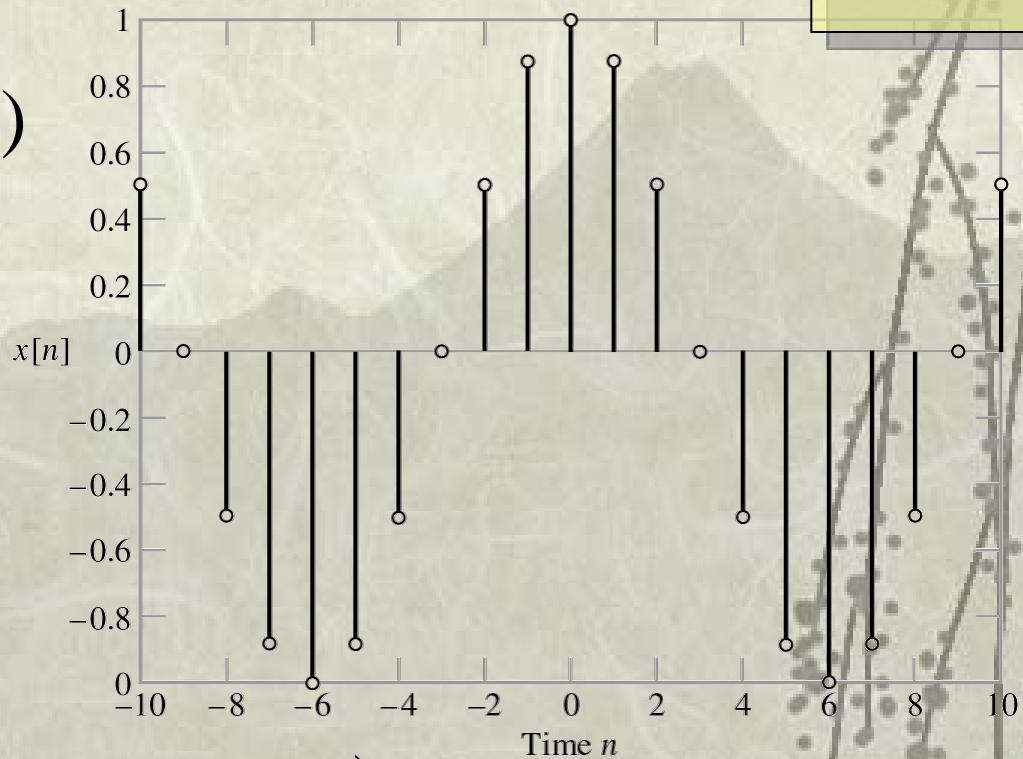


振盪電路之  
自然頻率

# 離散弦波訊號

ee: P37,38

$$x[n] = A \cos(\Omega n + \phi)$$



$$x[n + N] = A \cos(\Omega(n + N) + \phi)$$

$$\Rightarrow \Omega N = 2\pi m \quad \text{radians,}$$

or 
$$\Omega = \frac{2\pi m}{N} \quad \text{rad/cycle, } m, N \in \text{整數}$$

# 基本三角恆等式

$$\cos^2(\theta) = \frac{1}{2}(1 + \cos(2\theta))$$

$$\sin^2(\theta) = \frac{1}{2}(1 - \cos(2\theta))$$

$$\cos(2\theta) = 2\cos^2(\theta) - 1 = 1 - 2\sin^2(\theta)$$

$$\sin(2\theta) = 2\sin(\theta)\cos(\theta)$$

$$\sin(\theta \pm \phi) = \sin(\theta)\cos(\phi) \pm \cos(\theta)\sin(\phi)$$

$$\cos(\theta \pm \phi) = \cos(\theta)\cos(\phi) \mp \sin(\theta)\sin(\phi)$$

$$\sin(\theta)\sin(\phi) = \frac{1}{2}[\cos(\theta - \phi) - \cos(\theta + \phi)]$$

$$\cos(\theta)\cos(\phi) = \frac{1}{2}[\cos(\theta - \phi) + \cos(\theta + \phi)]$$

$$\sin(\theta)\cos(\phi) = \frac{1}{2}[\sin(\theta - \phi) + \sin(\theta + \phi)]$$

## Ex 1.7

$$x_1[n] = \sin(5\pi n), \quad x_2[n] = \sqrt{3} \cos(5\pi n)$$

(a) 求主週期(fundamental period)

$$(b) \text{求 } y[n] = x_1[n] + x_2[n] = A \cos(\Omega n + \phi)$$

$$\Omega N = 2\pi m, \quad \Omega = 5\pi$$

$$N = \frac{2\pi m}{\Omega} \xrightarrow{\text{最小整數}} N = 2$$

$$y[n] = x_1[n] + x_2[n] = \sin(5\pi n) + \sqrt{3} \cos(5\pi n)$$

$$y[n] = A \cos(\Omega n + \phi) = A[\cos(\Omega n) \cos(\phi) - \sin(\Omega n) \sin(\phi)]$$

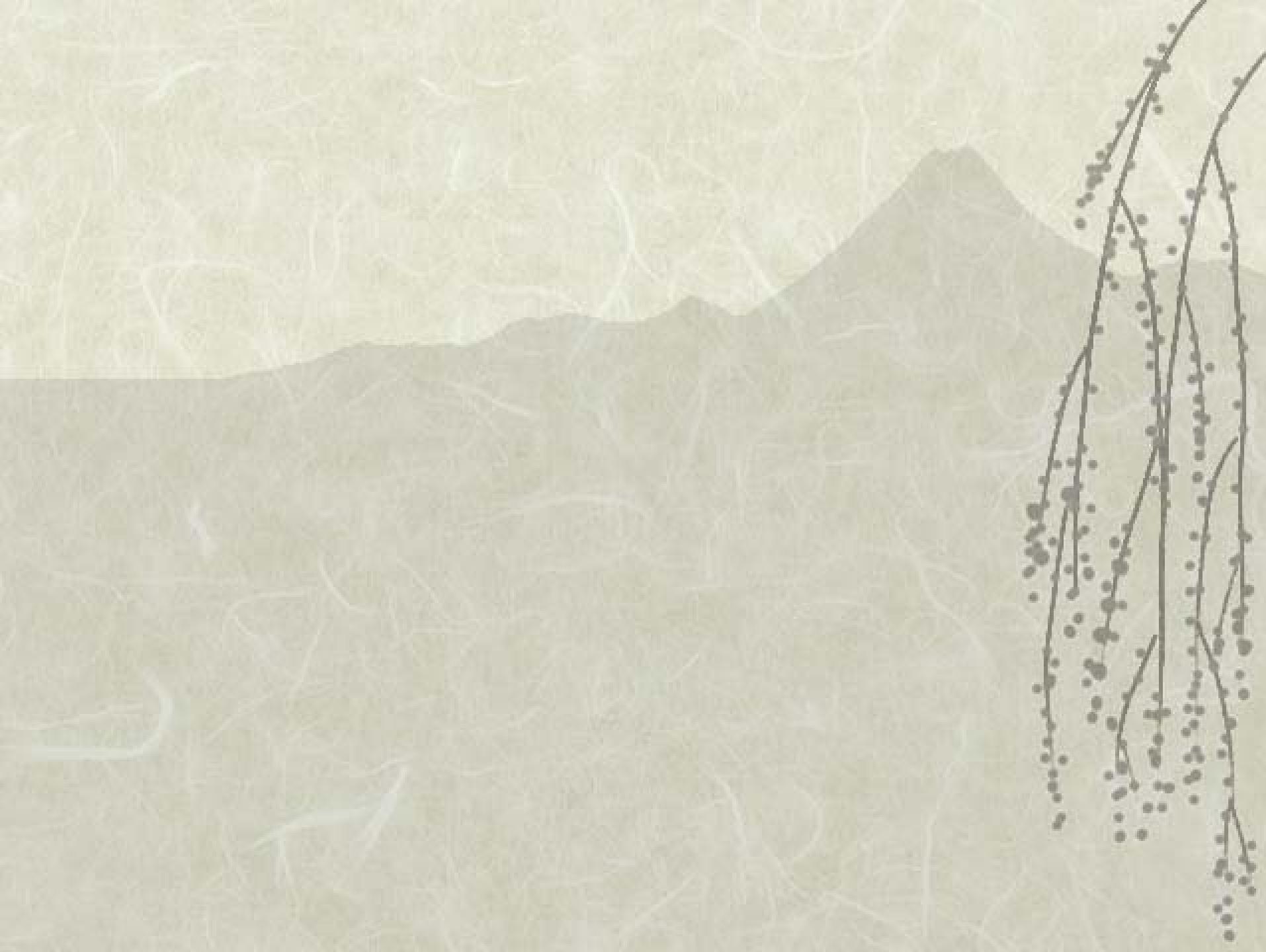
$$\rightarrow \Omega = 5\pi,$$

$$\rightarrow A \sin(\phi) = -1,$$

$$\rightarrow A \cos(\phi) = \sqrt{3}$$

$$\rightarrow \frac{A \sin(\phi)}{A \cos(\phi)} = \tan(\phi) = \frac{-1}{\sqrt{3}}, \therefore \phi = -\pi/3$$

$$A \sin(\phi) = -1 \rightarrow A = -1 / \sin(-\pi/3) = 2$$



# 弦波訊號與複數指數訊號

*Euler's identity*

$$e^{j\theta} = \cos(\theta) + j \sin(\theta)$$

複數指數訊號

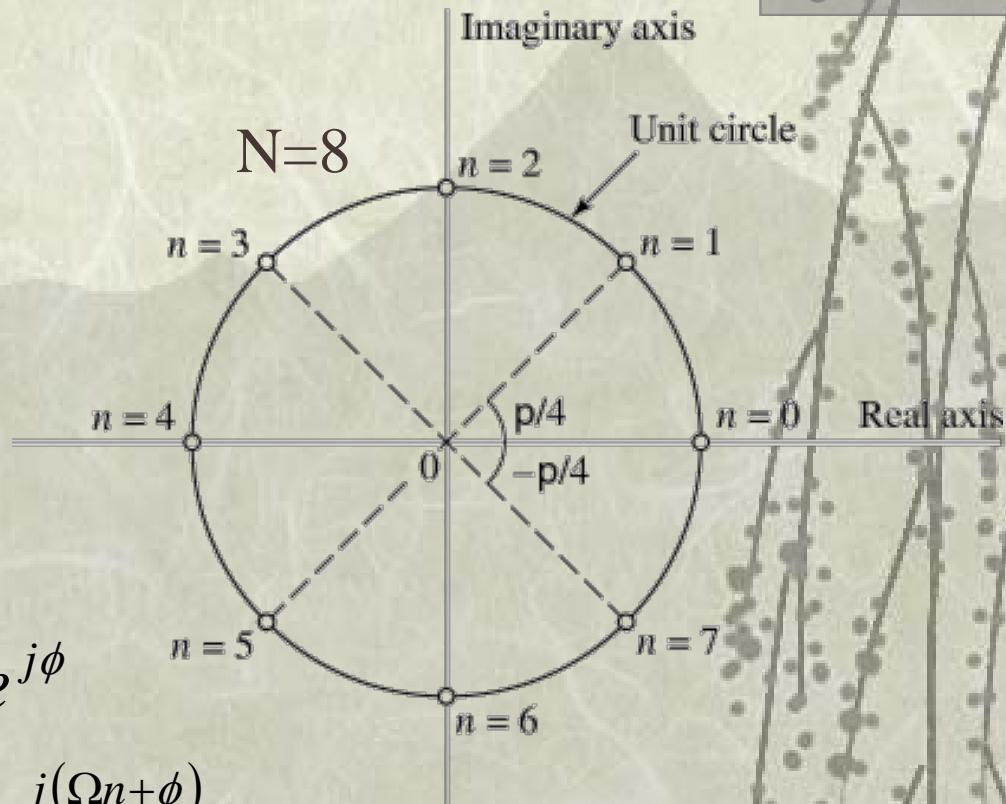
$$Be^{j\varpi t}, \quad \text{其中 } B = Ae^{j\phi}$$

$$Be^{j\varpi t} = Ae^{j\phi} e^{j\varpi t} = Ae^{j(\varpi t + \phi)}$$

$$\xrightarrow{\text{取實部}} A \cos(\varpi t + \phi) = \operatorname{Re}\{Be^{j\varpi t}\}$$

$$\xrightarrow{\text{取虛部}} A \sin(\varpi t + \phi) = \operatorname{Im}\{Be^{j\varpi t}\}$$

# 離散弦波訊號與複數指數訊號



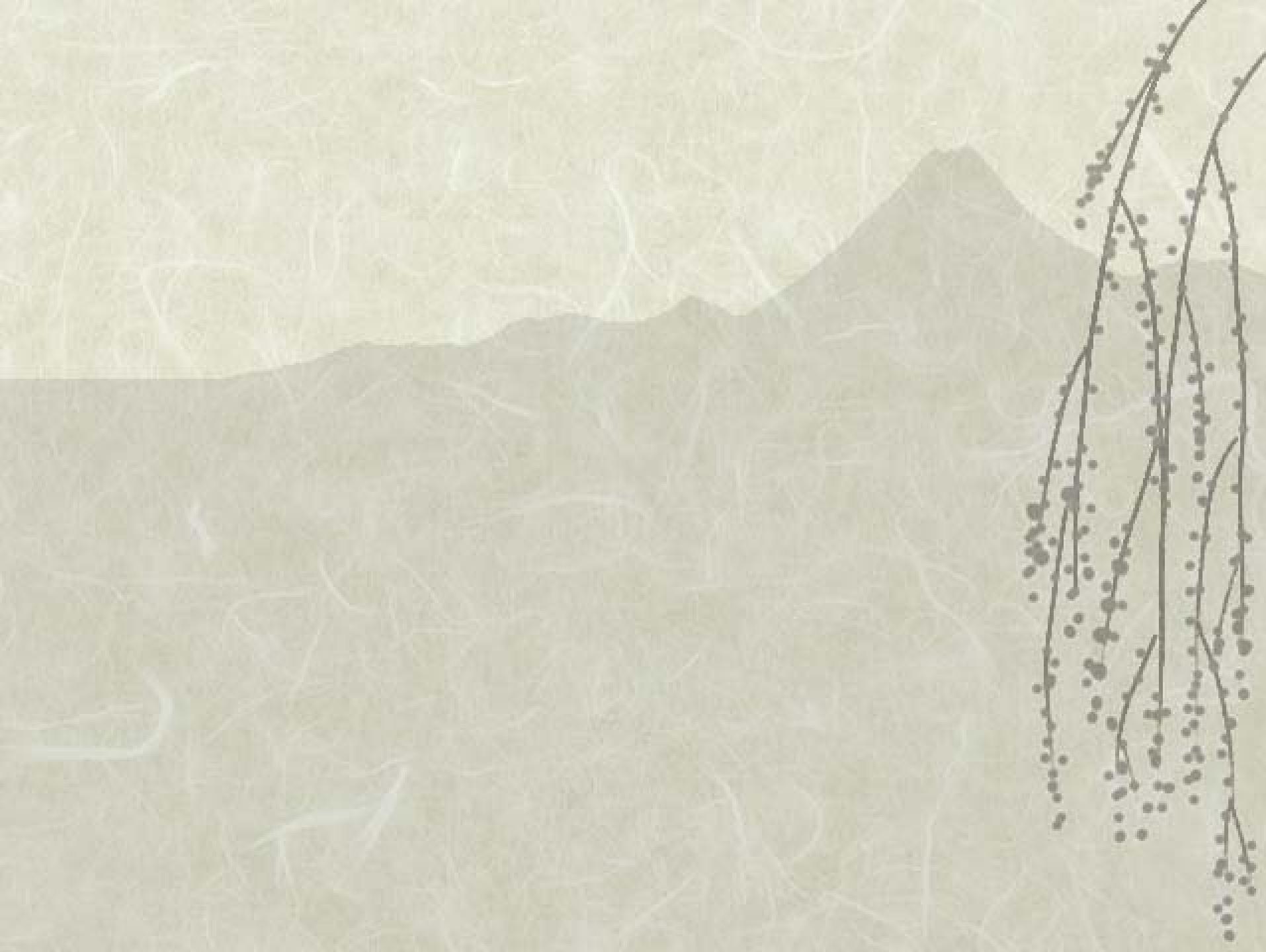
## 複數指數訊號

$$Be^{j\Omega n}, \quad \text{其中 } B = Ae^{j\phi}$$

$$Be^{j\Omega n} = Ae^{j\phi} e^{j\Omega n} = Ae^{j(\Omega n + \phi)}$$

取實部  $\rightarrow A \cos(\Omega n + \phi) = \operatorname{Re}\{Be^{j\Omega n}\}$

取虛部  $\rightarrow A \sin(\Omega n + \phi) = \operatorname{Im}\{Be^{j\Omega n}\}$



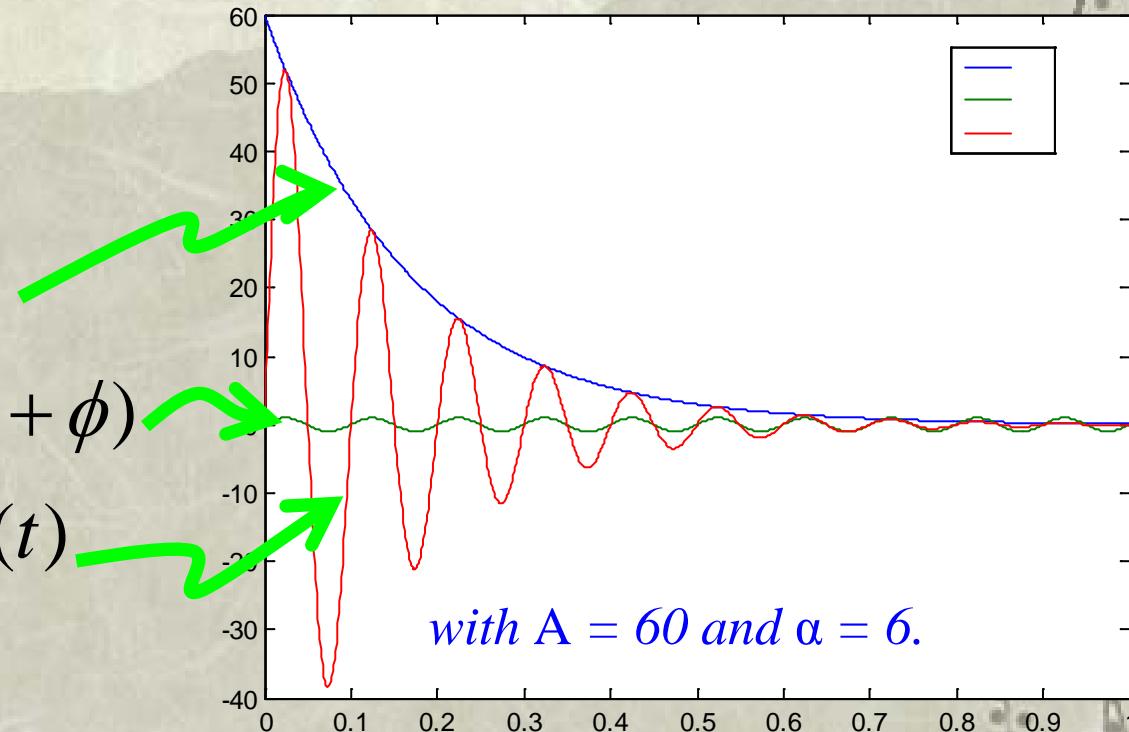
# 指數衰退之弦波訊號

$$x(t) = Ae^{-at} \sin(\varpi t + \phi)$$

$$x_1(t) = Ae^{-at}$$

$$x_2(t) = \sin(\varpi t + \phi)$$

$$x(t) = x_1(t)x_2(t)$$



# 計算: RLC 電路

KCL:

$$C \frac{d}{dt} v(t) + \frac{1}{R} v(t) + \frac{1}{L} \int_{-\infty}^t v(\tau) d\tau = 0$$

$$C \frac{d^2}{dt^2} v(t) + \frac{1}{R} \frac{d}{dt} v(t) + \frac{1}{L} v(t) = 0$$

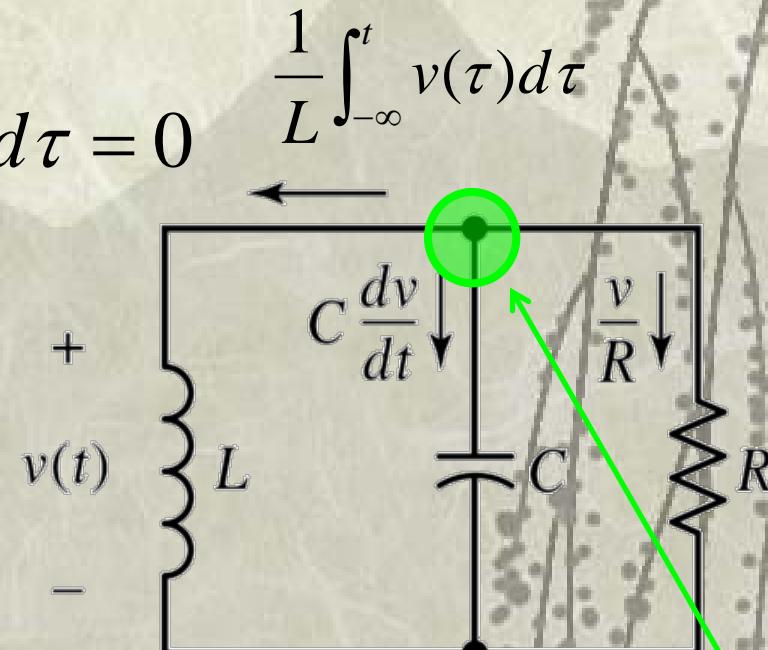
假設解  $v(t) = A e^{(\sigma+j\omega)t}$

$$RLC(\sigma + j\omega)^2 + L(\sigma + j\omega) + R = 0$$

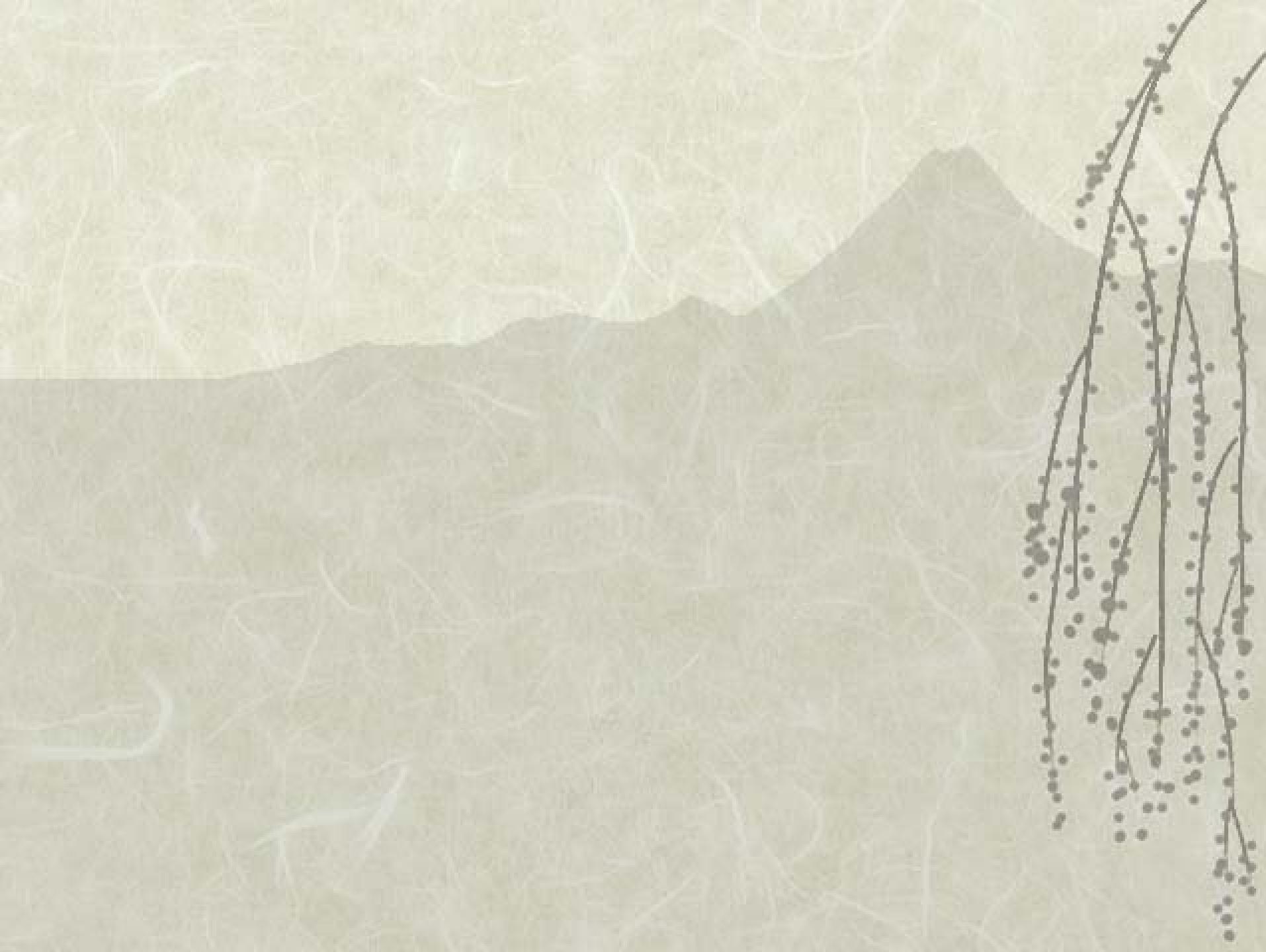
$$\rightarrow \sigma + j\omega = \frac{-L \pm \sqrt{L^2 - 4R^2LC}}{2RLC} = \frac{-1}{2RC} \pm j\sqrt{\frac{1}{LC} - \frac{1}{4R^2C^2}}$$

$$\therefore \sigma = \frac{-1}{2RC}, \quad \omega = \sqrt{\frac{1}{LC} - \frac{1}{4R^2C^2}} \quad \text{assume} \quad R > \sqrt{L/(4C)}$$

初值  $v(0) = Ae^0 = A$



節點總流出電流=0

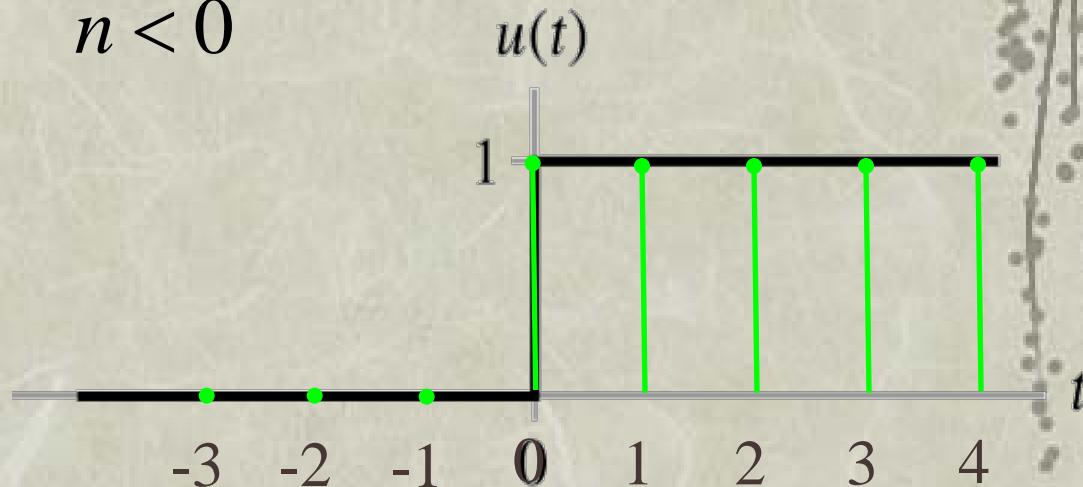


# 單位步階訊號, unit-step signal

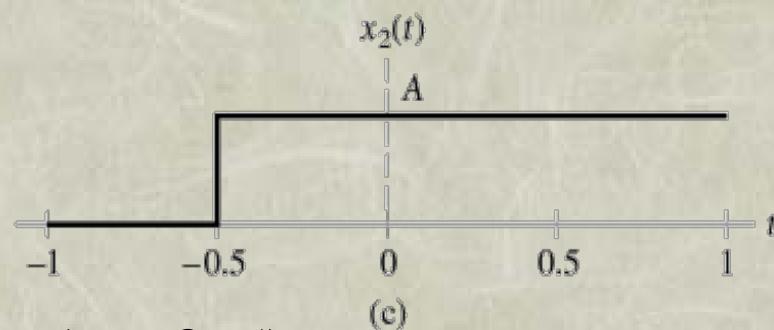
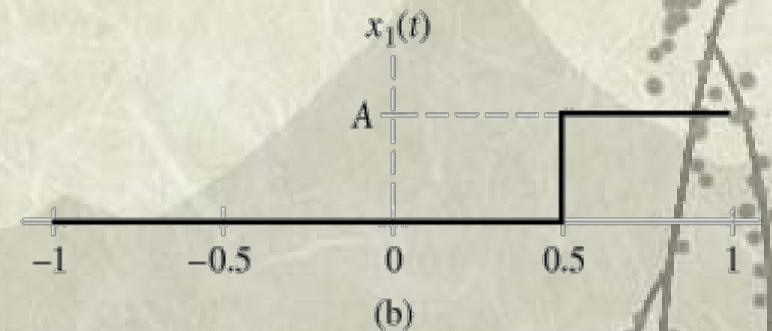
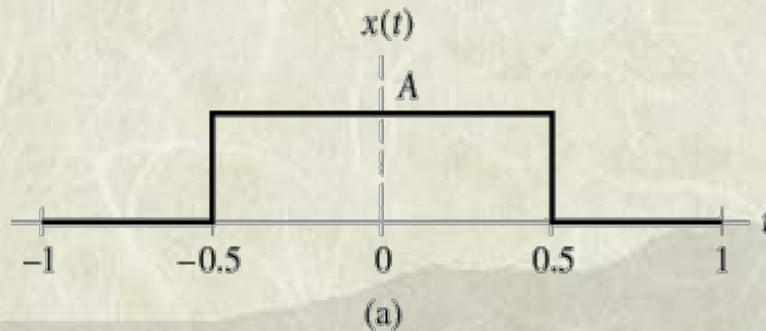
ee: P43  
Fig 1.38

$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

$$u[n] = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$$



以單位步階訊號表示下列訊號



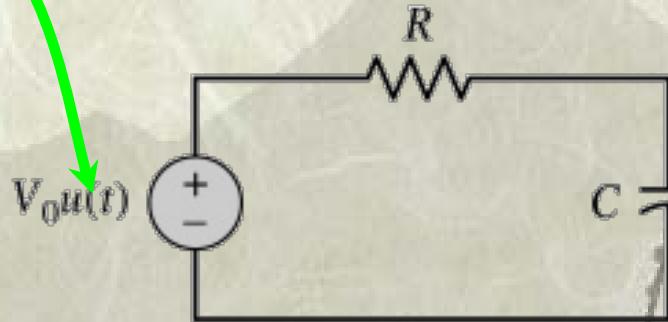
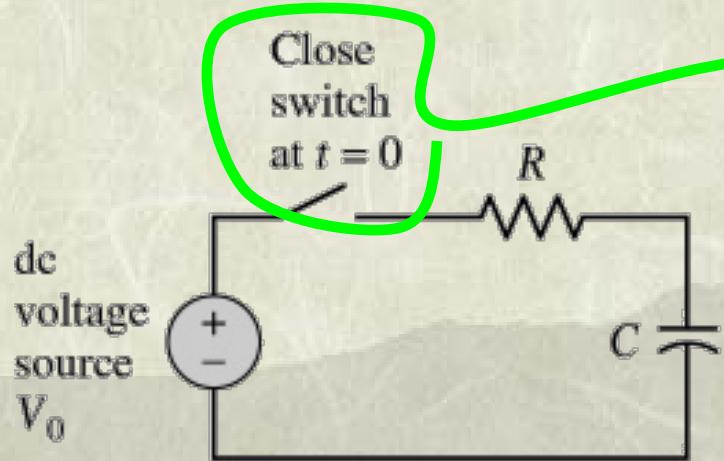
$$x_1(t) = u(t - 0.5)$$

$$x_2(t) = u(t + 0.5) = u(t - (-0.5))$$

$$\therefore x(t) = x_2(t) - x_1(t) = u(t + 0.5) - u(t - 0.5)$$

# 以步階訊號描述開關動作

ee: P45  
Fig 1.40



(a)

$$v(t) = v_p + Be^{at}, \quad v_p \text{ 為特解}$$

(b)

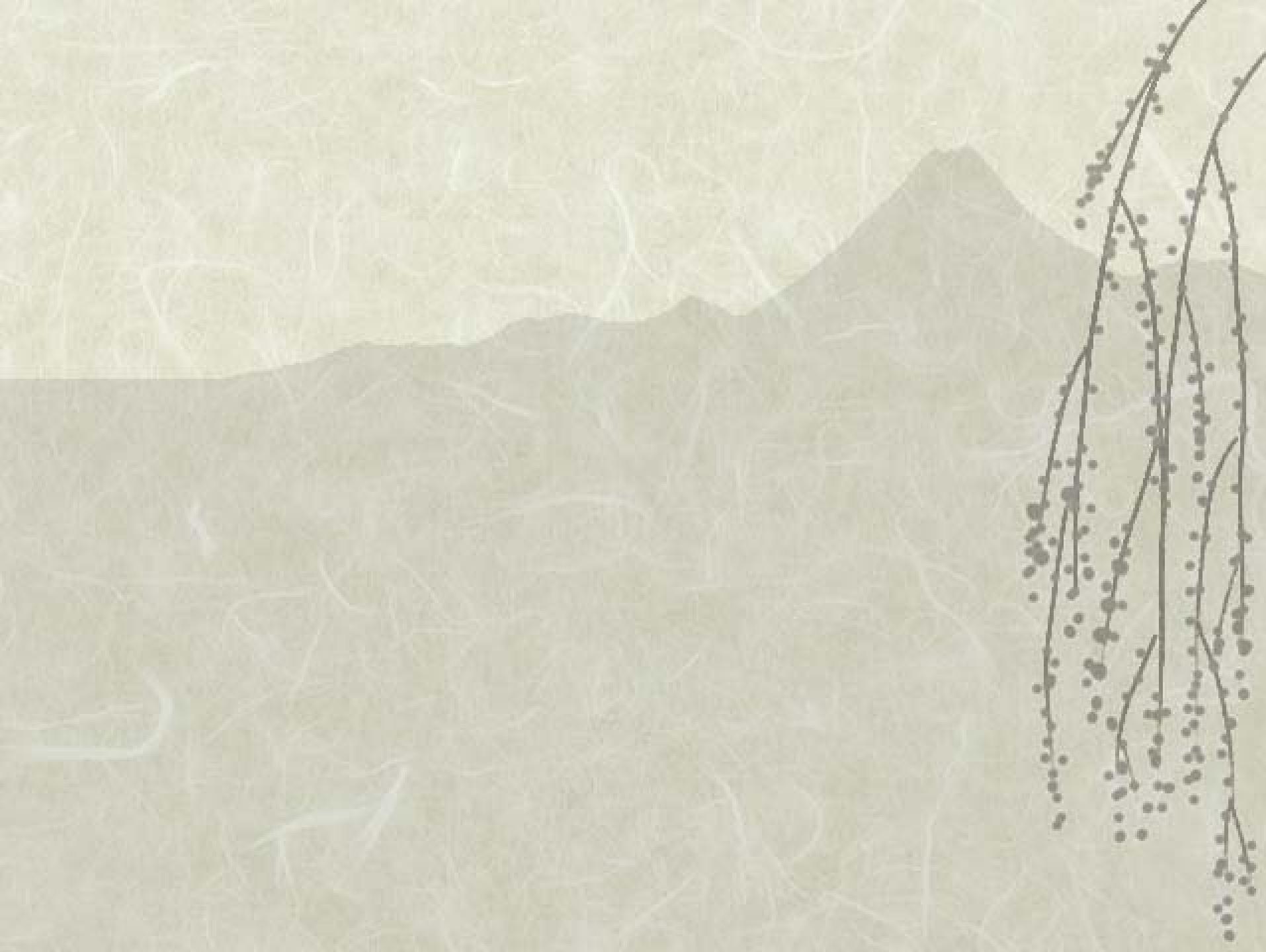
$$a = -1/(RC)$$

$$\because v(0) = 0, v(\infty) = V_0$$

$$\rightarrow v_p + B = 0, \quad V_0 = v_p \rightarrow B = -v_p$$

$$v(t) = V_0(1 - e^{-t/(RC)})u(t)$$

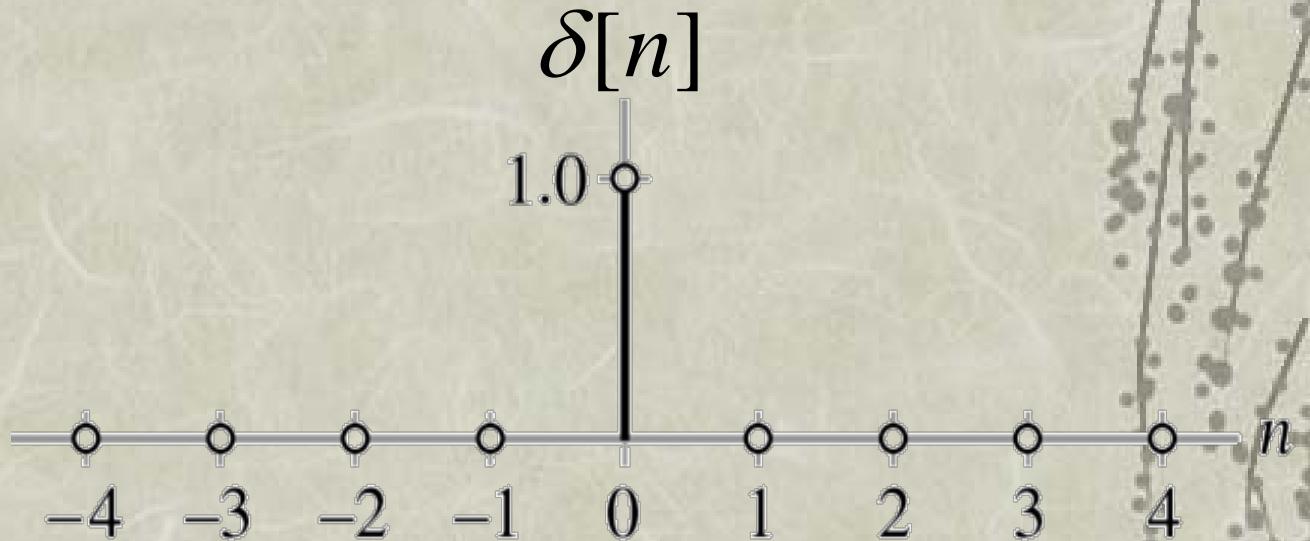
描述開關動作  
後之輸出



# 離散脈衝(Discrete-time form of impulse.)

ee: P46  
Fig 1.41

$$\delta[n] = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$$

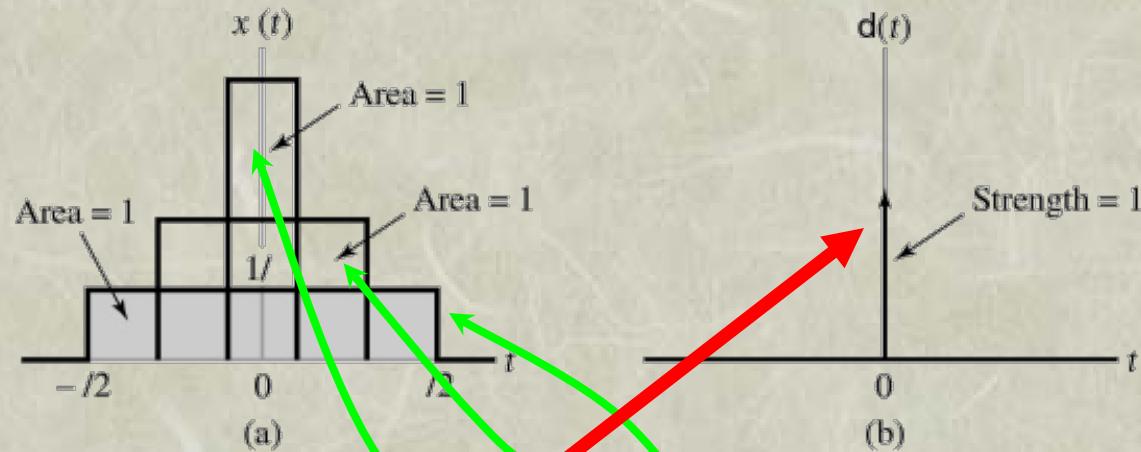


# 脈衝(Dirac delta)

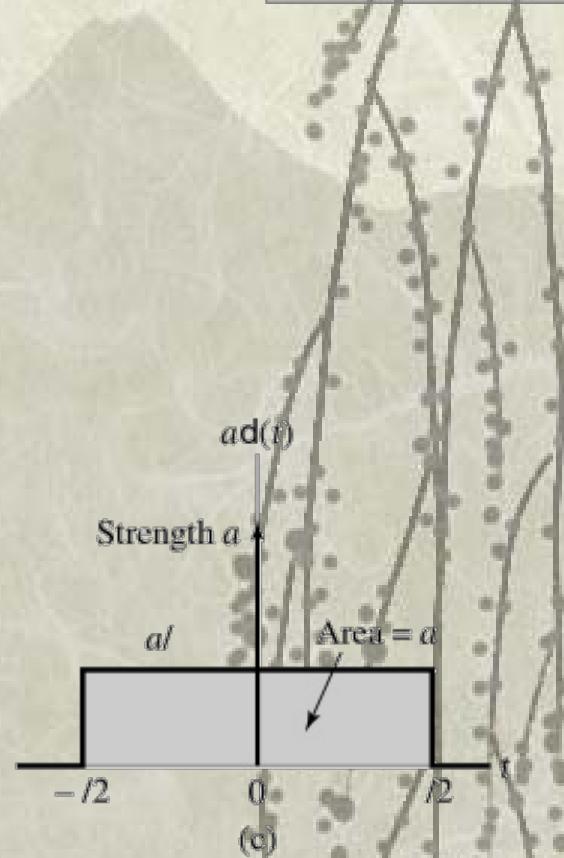
$$\delta(t) = 0, \quad \text{for } t \neq 0$$

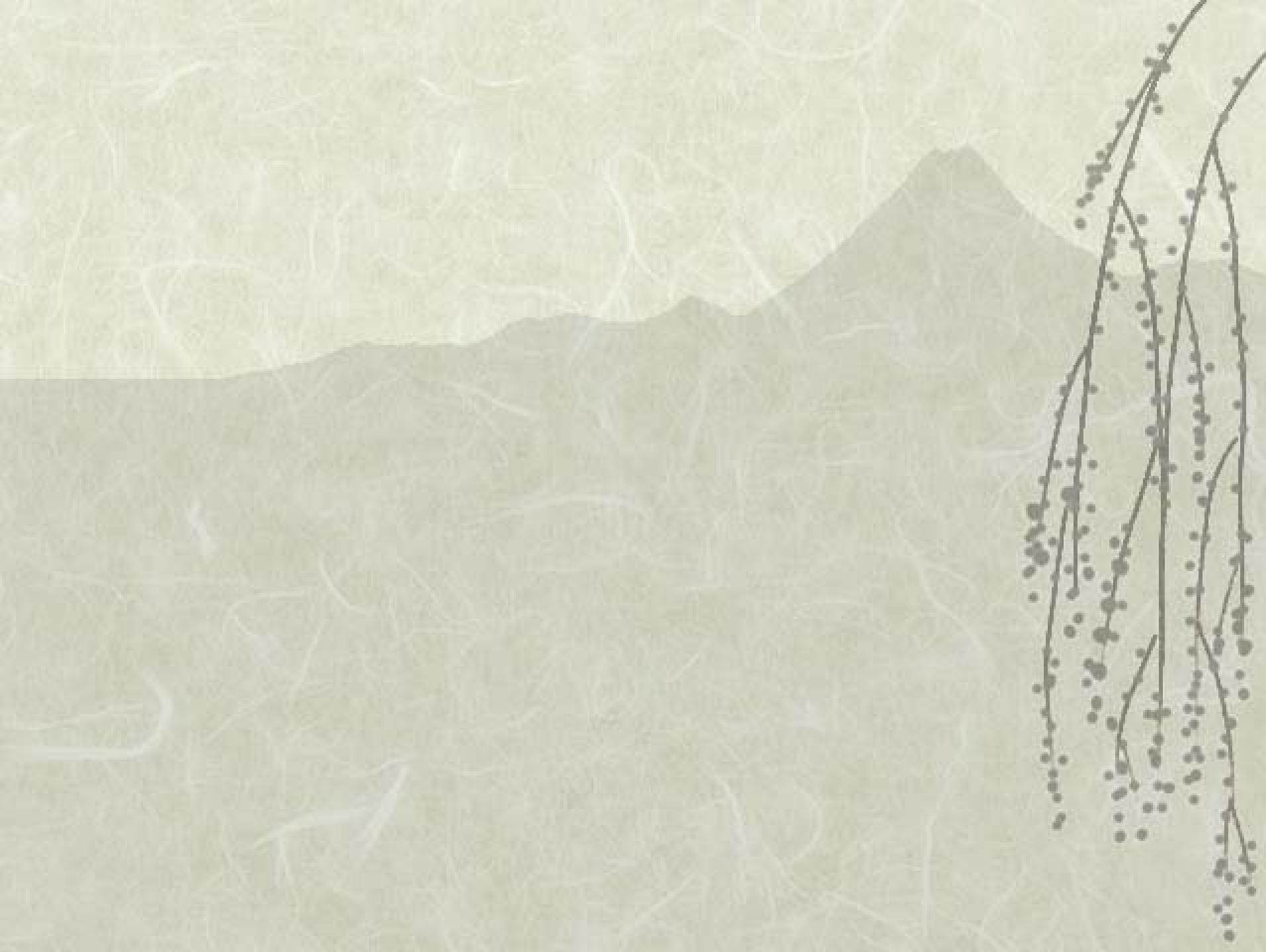
and

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$



$$\delta(t) = \lim_{\Delta \rightarrow 0} x_{\Delta}(t)$$





# Dirac delta & unit step

$$\delta(t) = \frac{d}{dt} u(t)$$

$$u(t) = \int_{-\infty}^t \delta(\tau) d\tau$$

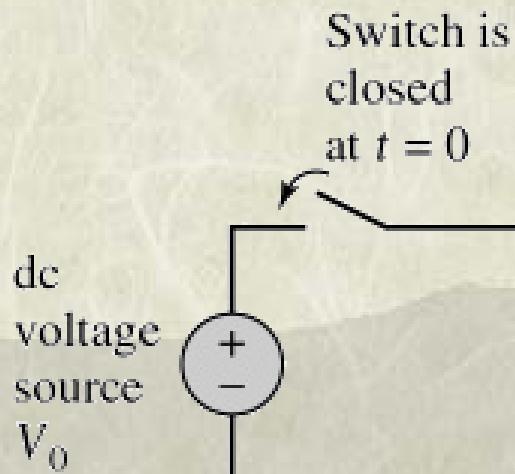
$$\delta(t) = 0, \quad \text{for } t \neq 0$$

and

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$



$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

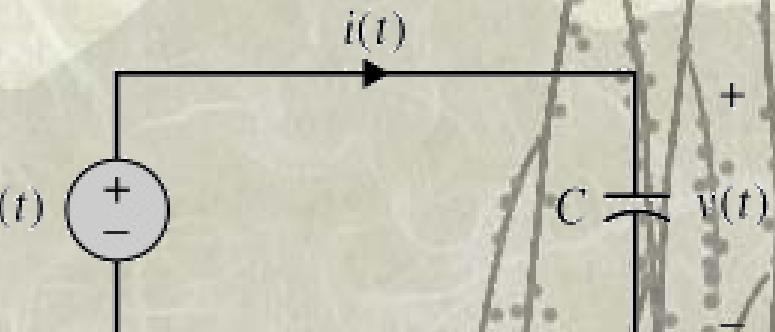
**Ex 1.10**

(a)

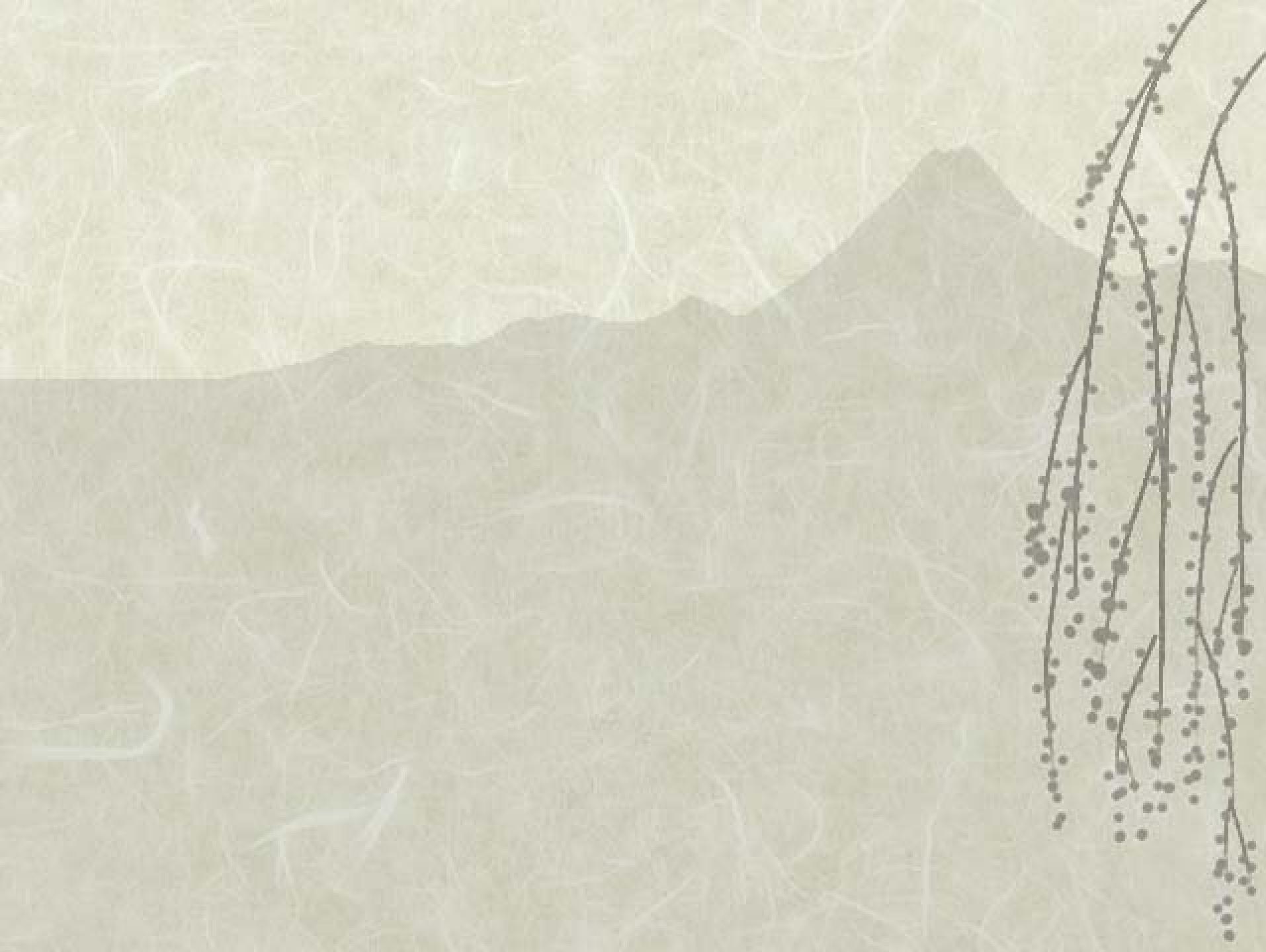
$$i(t) = C \frac{dv(t)}{dt}$$

$$\because v(t) = V_0 u(t)$$

$$\therefore i(t) = CV_0 \delta(t)$$



(b)



# Dirac delta 之時間變數運算

shifting & 積分：◦◦◦

$$\int_{-\infty}^{\infty} x(t) \delta(t - t_0) = x(t - t_0)$$

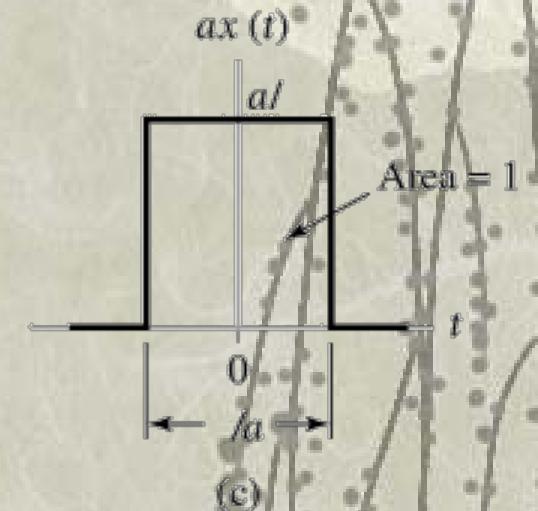
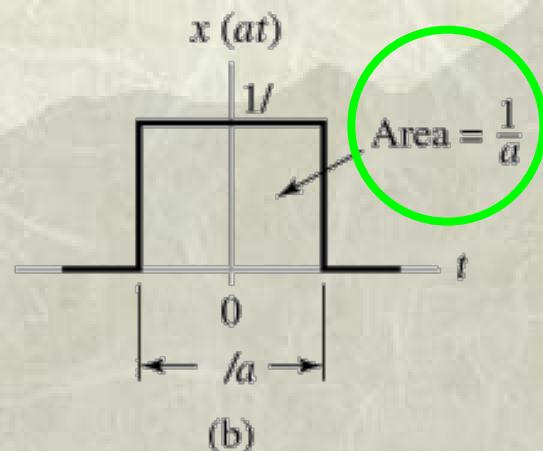
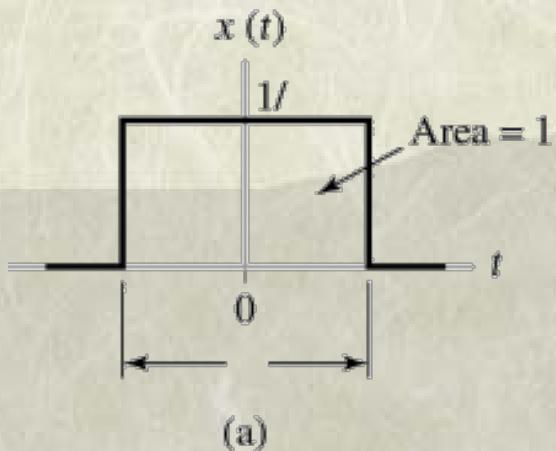
Dirac delta 可  
視為一 shift  
operator

scaling :

$$\delta(at) = \frac{1}{a} \delta(t), a > 0$$

# 圖解脈衝訊號之scaling 運算

ee: P48  
Fig. 1.44

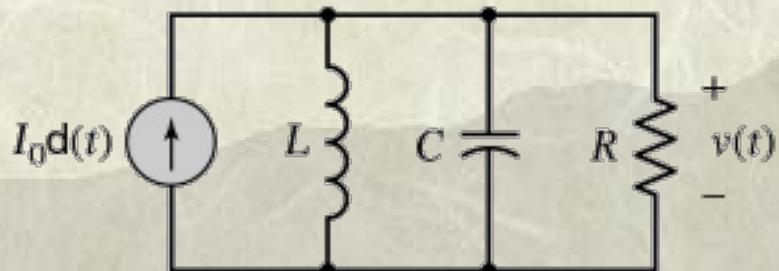


$$\delta(at) = \lim_{\Delta \rightarrow 0} x_\Delta(at) = \lim_{\Delta \rightarrow 0} \frac{1}{a} x_\Delta(t)$$

$$\delta(at) = \frac{1}{a} \delta(t)$$

# Parallel LRC

(a) find the value of the voltage  $v(t)$  at  $t = 0^+$



(a)

$$d(t) = \delta(t)$$

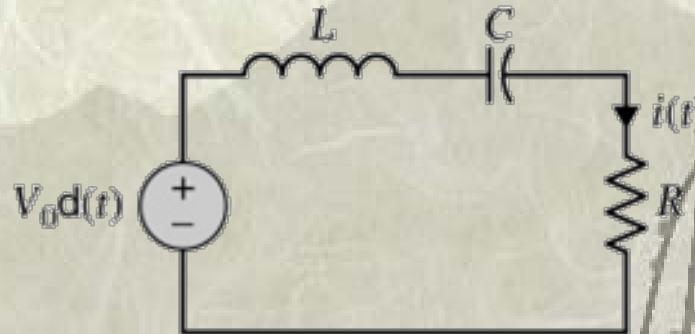
$$(a) I_0 \delta(t) = \frac{1}{L} \int_{-\infty}^t v(t) dt + C \frac{d}{dt} v(t) + \frac{1}{R} v(t)$$

$$\xrightarrow{\text{兩邊積分至 } 0^+} \int_{-\infty}^{0^+} I_0 \delta(t) dt = C v(0^+) \quad , \therefore V_0 = I_0 / C$$

# P1.23 series LCR

(a) find the value of the current  $i(t)$  at  $t = 0^+$

(b) write the integro - differential equation of  $i(t)$  for  $t \geq 0^+$



$$d(t) = \delta(t)$$

$$(a) V_0 \delta(t) = L \frac{d}{dt} i(t) + \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau + R i(t) \quad (b)$$

$$\xrightarrow{\text{兩邊積分至 } 0^+} \int_{-\infty}^{0^+} V_0 \delta(t) dt = L i(0^+) \quad , \therefore I_0 = V_0 / L$$

$$(b) V_0 \delta(t) = L \frac{d}{dt} i(t) + \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau + R i(t)$$

$$\xrightarrow{t \geq 0^+} L \frac{d}{dt} i(t) + \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau + R i(t) = 0$$

# Derivatives of the Impulse

$$\delta^{(1)}(t) = \frac{d\delta(t)}{dt} = \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} [\delta(t + \Delta/2) - \delta(t - \Delta/2)]$$

doublet性質

$$\int_{-\infty}^{\infty} \delta^{(1)}(t) dt = 0 \Rightarrow \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} \left[ \int_{-\infty}^{\infty} \delta(t + \Delta/2) dt - \int_{-\infty}^{\infty} \delta(t - \Delta/2) dt \right] = 0$$

$$\int_{-\infty}^{\infty} f(t) \delta^{(1)}(t - t_0) dt = \frac{d}{dt} f(t) \Big|_{t=t_0}$$

➡

$$= \int_{-\infty}^{\infty} f(t) \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} [\delta(t - t_0 + \Delta/2) - \delta(t - t_0 - \Delta/2)] dt$$

doublet微分

$$\frac{d}{dt} \delta^{(1)}(t) = \frac{d^2}{dt^2} \delta(t) = \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} [f(t - t_0 + \Delta/2) - f(t - t_0 - \Delta/2)] = \frac{d}{dt} f(t) \Big|_{t=t_0}$$

$$= \lim_{\Delta \rightarrow 0} \frac{\delta^{(1)}(t + \Delta/2) - \delta^{(1)}(t - \Delta/2)}{\Delta}$$

# P1.24

$$\delta^{(2)}(t) = \frac{d}{dt} \delta^{(1)}(t) = \lim_{\Delta \rightarrow 0} \frac{\delta^{(1)}(t + \Delta/2) - \delta^{(1)}(t - \Delta/2)}{\Delta}$$

$\delta^{(2)}(t)$ 性質

$$(a) \int_{-\infty}^{\infty} f(t) \delta^{(2)}(t - t_0) dt$$

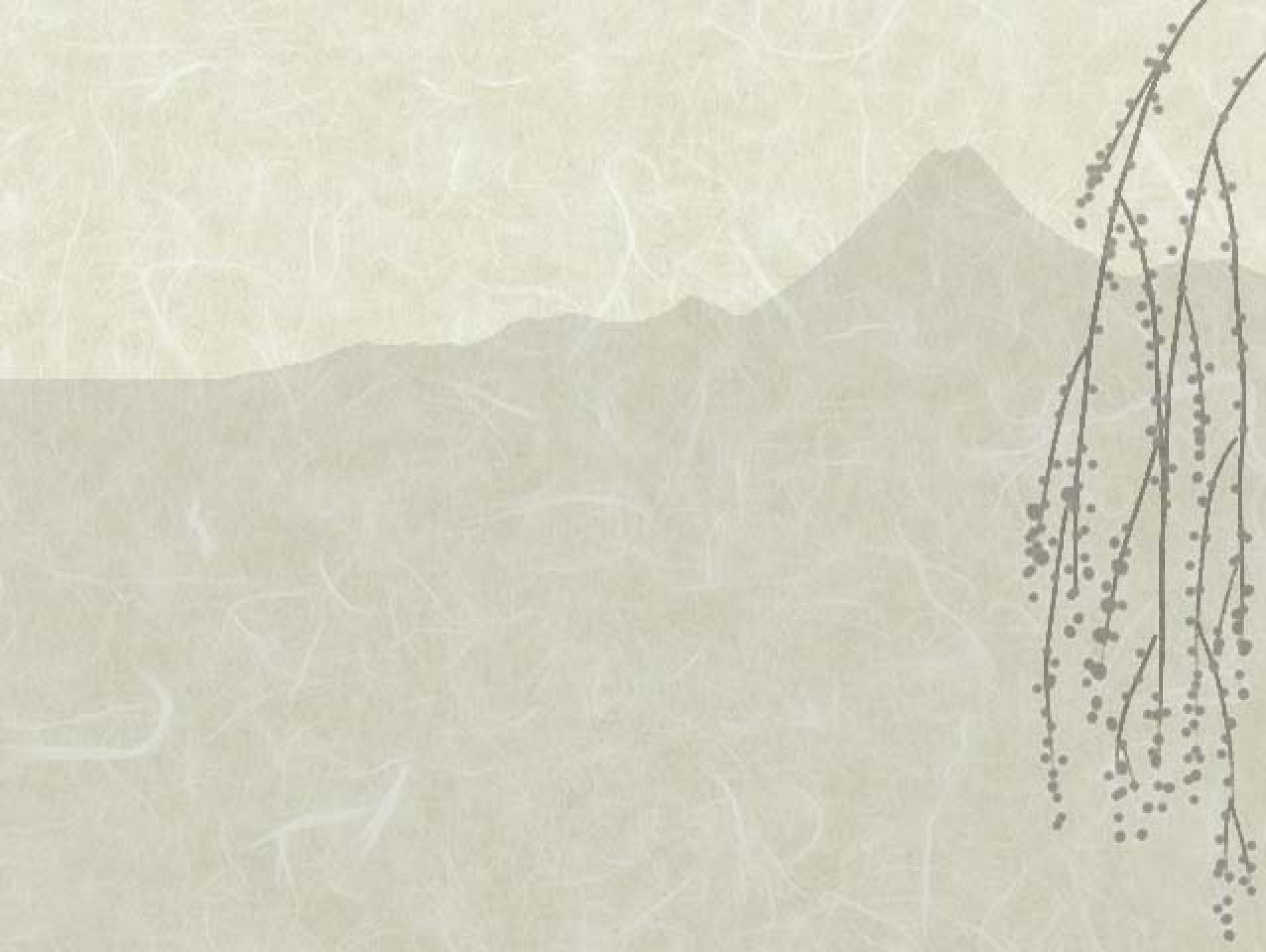
$$= \lim_{\Delta \rightarrow 0} \frac{\int_{-\infty}^{\infty} f(t) \delta^{(1)}(t - t_0 + \Delta/2) dt - \int_{-\infty}^{\infty} f(t) \delta^{(1)}(t - t_0 - \Delta/2) dt}{\Delta}$$

$$= \lim_{\Delta \rightarrow 0} \frac{\frac{d}{d\tau} f(\tau - t_0) \Big|_{\tau=t+\Delta/2} - \frac{d}{d\tau} f(\tau - t_0) \Big|_{\tau=t-\Delta/2}}{\Delta} = \frac{d}{dt} \left( \frac{d}{dt} f(t) \right) \Big|_{t=t_0}$$

$$(b) \int_{-\infty}^{\infty} f(t) \delta^{(n)}(t - t_0) dt$$

$$= \lim_{\Delta \rightarrow 0} \frac{\int_{-\infty}^{\infty} f(t) \delta^{(n-1)}(t - t_0 + \Delta/2) dt - \int_{-\infty}^{\infty} f(t) \delta^{(n-1)}(t - t_0 - \Delta/2) dt}{\Delta}$$

$$= \dots = \frac{d^n}{dt^n} f(t) \Big|_{t=t_0}$$

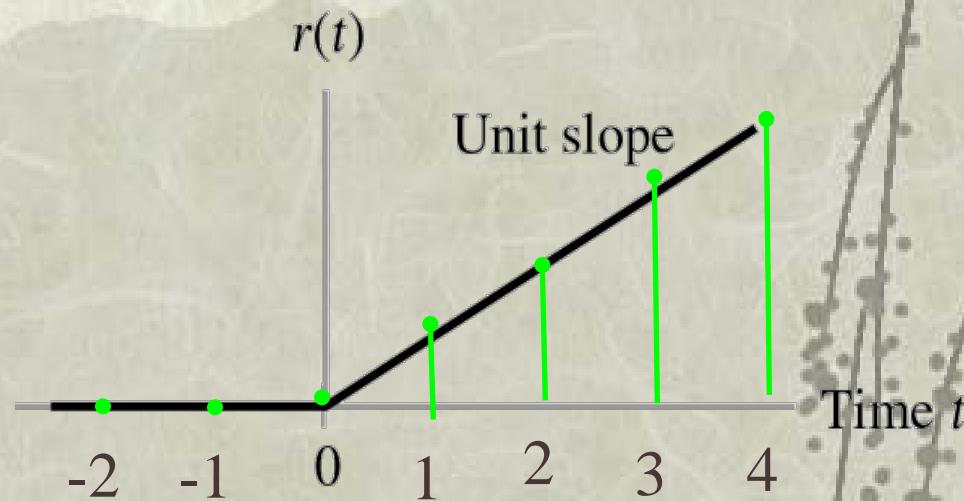


# Ramp Signal

ee: P51,52

Fig. 1.46,47

$$r(t) = \begin{cases} t, & t \geq 0 \\ 0, & t < 0 \end{cases} = tu(t)$$



$$r[n] = \begin{cases} n, & n \geq 0 \\ 0, & n < 0 \end{cases} = nu[n]$$

# Ex 1.11

$$i(t) = I_0 u(t)$$

$$v(t) = \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau$$

$$= \frac{1}{C} \int_{-\infty}^t I_0 u(\tau) d\tau$$

$$= \frac{I_0}{C} r(t)$$

dc  
current  
source  
 $I_0$

